

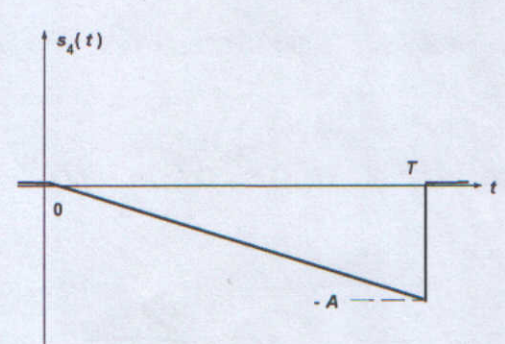
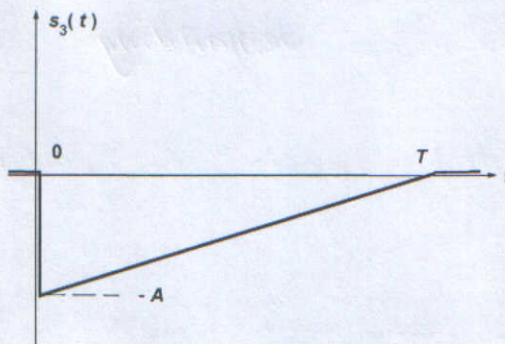
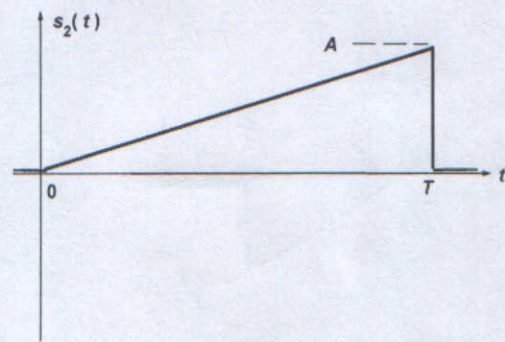
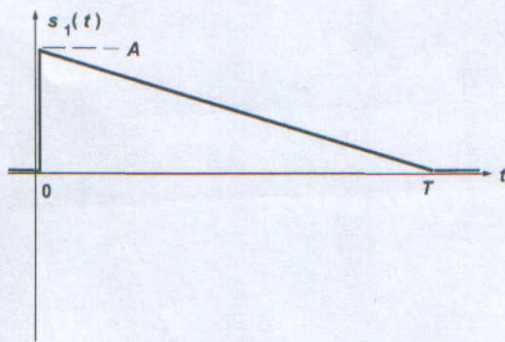
Çankaya University – ECE Department – ECE 376

Student Name :
Student Number :

Open source exam
Duration : 2 hours

Questions

1. (70 Points) The four signals $s_1(t)$, $s_2(t)$, $s_3(t)$ and $s_4(t)$ are given as below. Determine what type of modulation (i.e., ASK, PSK, QAM, FSK), these signals represent and the parameters M and N . Find set of $\psi_i(t)$ **orthogonal** basis functions for $s_1(t)$, $s_2(t)$, $s_3(t)$ and $s_4(t)$. Draw the signal constellation diagram, showing the position, the length of vectors s_1 , s_2 , s_3 and s_4 and the distance between them. Comment whether the spacings between s_1 , s_2 , s_3 and s_4 are optimum (that is equally distributed on the constellation diagram), if not suggest ways to improve it by indicating this optimum arrangement on the constellation diagram. Draw the receiver diagram as correlator and MF, find the output from the MF detector, when $s_1(t)$ was transmitted and compare it to the output obtained from the correlator.



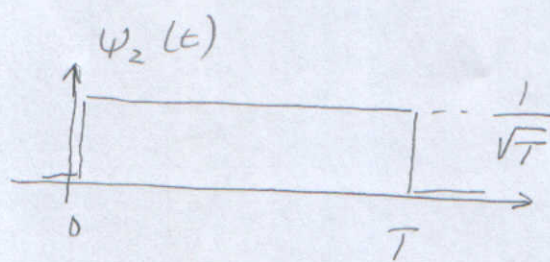
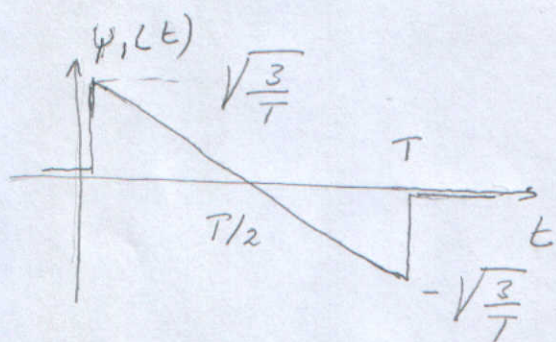
Note : Keep in mind that $\psi_i(t)$ basis functions must satisfy the **orthogonality** condition.

Solution : $s_1(t)$ and $s_2(t)$ can be written as follows;

$$s_1(t) = \frac{A}{T} (T-t) = A \left(1 - \frac{t}{T} \right), \quad s_2(t) = \frac{At}{T}$$

$$s_3(t) = -s_1(t), \quad s_4(t) = -s_2(t)$$

Because of the relations of $s_3(t)$ and $s_4(t)$ to $s_1(t)$ and $s_2(t)$, also keeping in mind that $s_1(t)$ and $s_2(t)$ cannot be expressed in terms of each other, so a minimum of two basis functions will be needed. By observing the rule of orthogonality, these may be as follows



The peak values are found by demanding that energies of $\psi_1(t)$ and $\psi_2(t)$ equate to unity. Thus

$$\psi_1(t) = \sqrt{\frac{3}{T}} \left(1 - \frac{2t}{T}\right), \quad \psi_2(t) = \frac{1}{\sqrt{T}} \quad 0 < t < T$$

It is easy to see from the above illustrations

$$\int_0^T \psi_1(t) \psi_2(t) dt = 0$$

Signals $s_1(t) \dots s_4(t)$ in terms of $\psi_1(t)$ and $\psi_2(t)$

$$s_1(t) = \frac{A}{2} \sqrt{\frac{T}{3}} \psi_1(t) + \frac{A}{2} \sqrt{T} \psi_2(t)$$

$$s_2(t) = -\frac{A}{2} \sqrt{\frac{T}{3}} \psi_1(t) + \frac{A}{2} \sqrt{T} \psi_2(t)$$

$$s_3(t) = -\frac{A}{2} \sqrt{\frac{T}{3}} \psi_1(t) - \frac{A}{2} \sqrt{T} \psi_2(t)$$

$$s_4(t) = \frac{A}{2} \sqrt{\frac{T}{3}} \psi_1(t) - \frac{A}{2} \sqrt{T} \psi_2(t)$$

The coefficients in front are the signal vector elements

Hence $\overbrace{s_{11}}$ $\overbrace{s_{12}}$

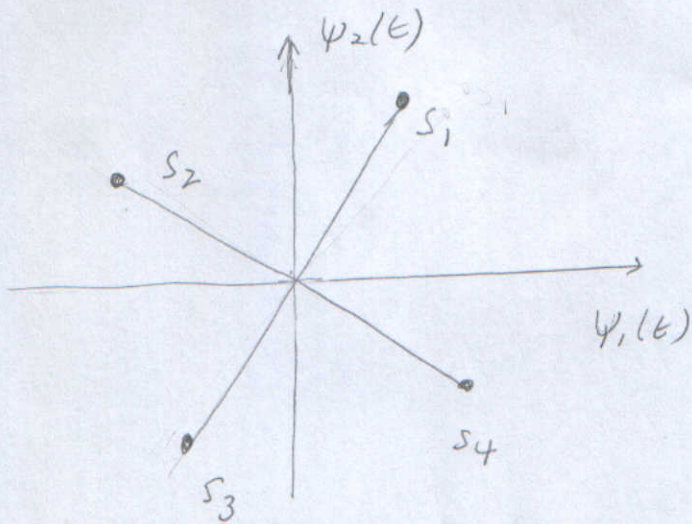
$$s_1 = \left[\frac{A}{2} \sqrt{\frac{T}{3}}, \frac{A}{2} \sqrt{T} \right], s_2 = \left[-\frac{A}{2} \sqrt{\frac{T}{3}}, \frac{A}{2} \sqrt{T} \right]$$

$$s_3 = \left[-\frac{A}{2} \sqrt{\frac{T}{3}}, -\frac{A}{2} \sqrt{T} \right], s_4 = \left[\frac{A}{2} \sqrt{\frac{T}{3}}, -\frac{A}{2} \sqrt{T} \right]$$

$$\left(\text{length of } s_1 \dots s_4 \right)^2 = \frac{A^2 T}{3}$$

$$= \text{Energies of } s_1(t), s_2(t) \\ s_3(t), s_4(t)$$

(2)

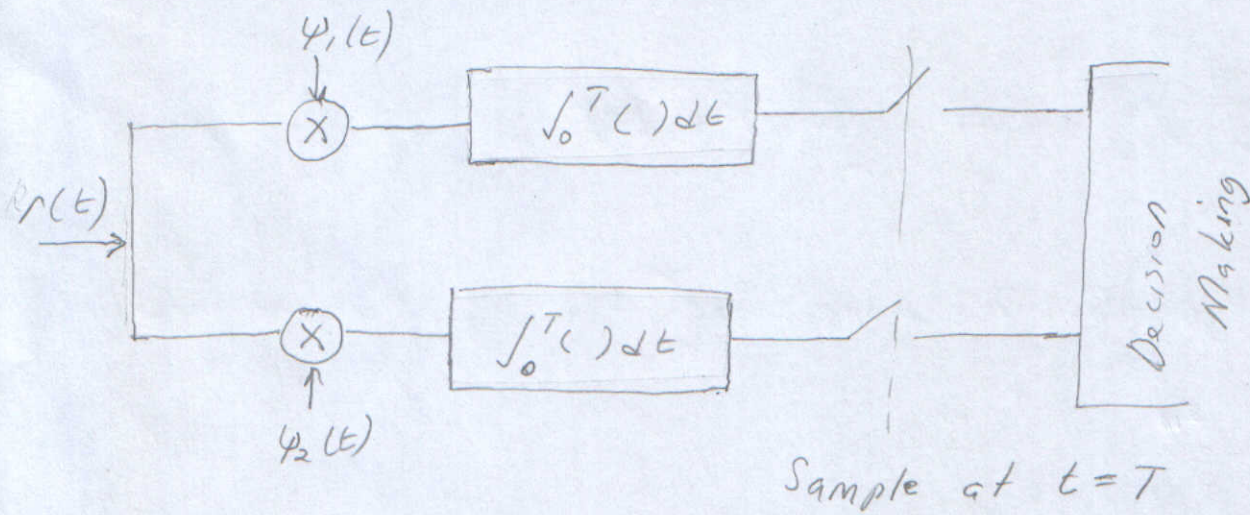


Constellation Diagram

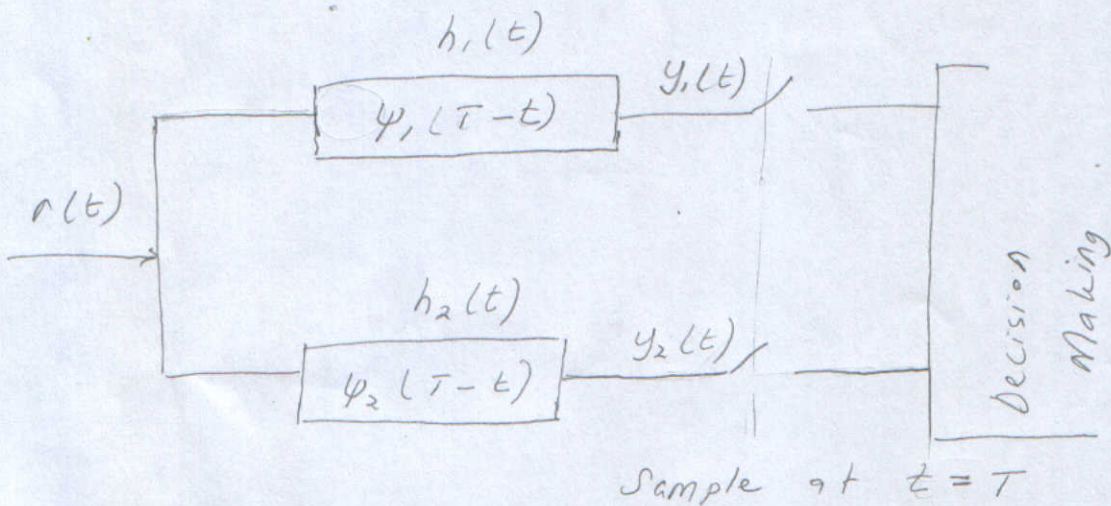
The spacing is optimum, since rotation of $\pm 15^\circ$ will give the well known 4 PSK or 4 QAM.

So optimum, no need for rearrangement

Receiver diagram as correlator



Receiver diagram as MF (Matched Filter)



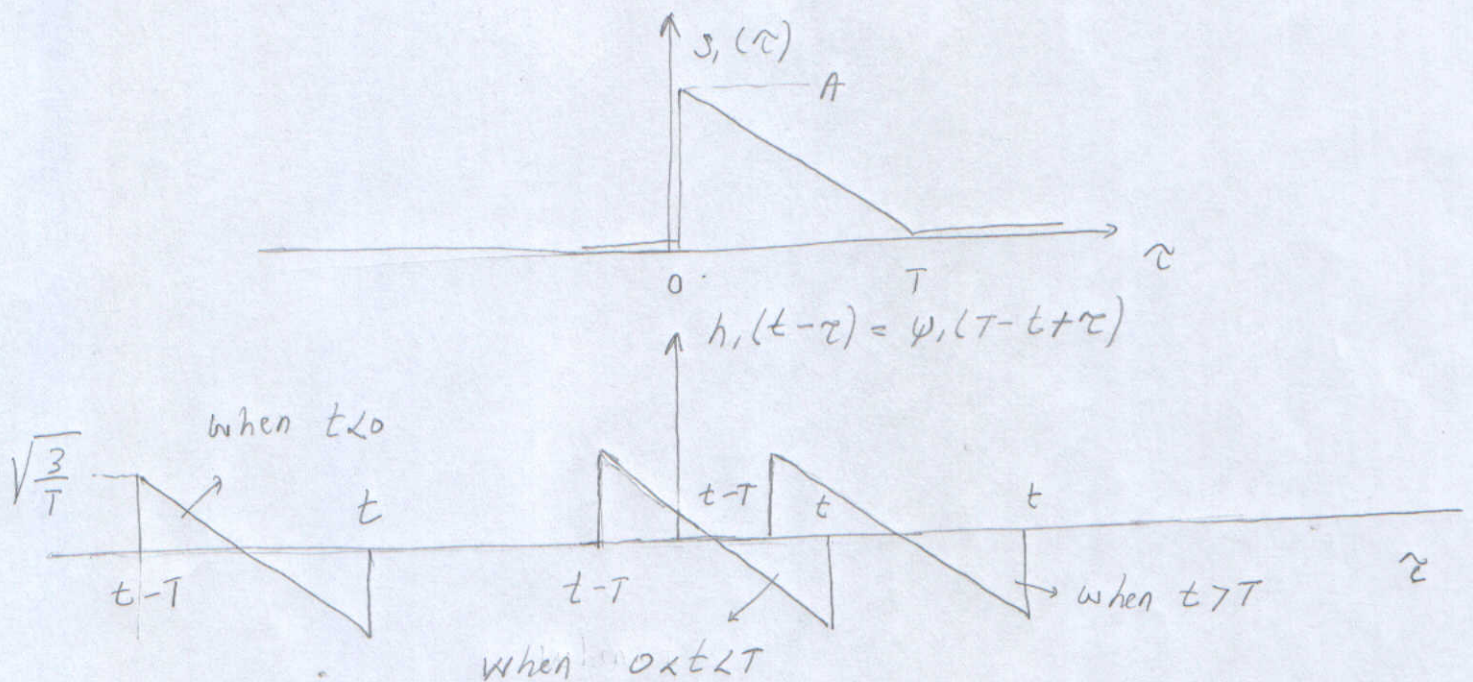
(3)

Taking the case of MF and $s_1(t)$ being transmitted, the convolution integrals and convolution graphs would be as follows;

$$y_1(t) = \int_0^T r(\tau) h_1(t-\tau) d\tau$$

$$= \int_0^T s_1(t) h_1(t-\tau) d\tau + \int_0^T n(t) h_1(t-\tau) d\tau$$

$$y_2(t) = \int_0^T r(\tau) h_2(t-\tau) d\tau$$



For upper branch of MF receiver at $0 < t < T$ (signal part only)
beginning of overlap

$$y_{11s_1}(t) = \int_0^t s_1(\tau) \psi_1(T-t+\tau) d\tau$$

$$y_{11s1}(t) = \frac{A}{T} \sqrt{\frac{3}{T}} \int_0^t \left[T - \tau - \frac{2(T^2 - T\tau + \tau t - \tau^2)}{T} \right] d\tau$$

$$= \frac{A}{2} \sqrt{\frac{3}{T}} \left(Tt - \frac{t^2}{2} - \frac{2T^2t - 2Tt^2 + t^3 - 2t^3/3}{T} \right)$$

More detailed steps are in EEM 467 MT2 solutions dated 11.12.2007.

Tests at $y_{11s1}(t=0) = 0$, $y_{11s1}(t=T) = \frac{A}{2} \sqrt{\frac{T}{3}}$: OK

At $T < t < 2T$ → end of overlap

$$y_{12s1}(t) = \int_{t-T}^T s_1(\tau) \psi_1(T-t+\tau) d\tau$$

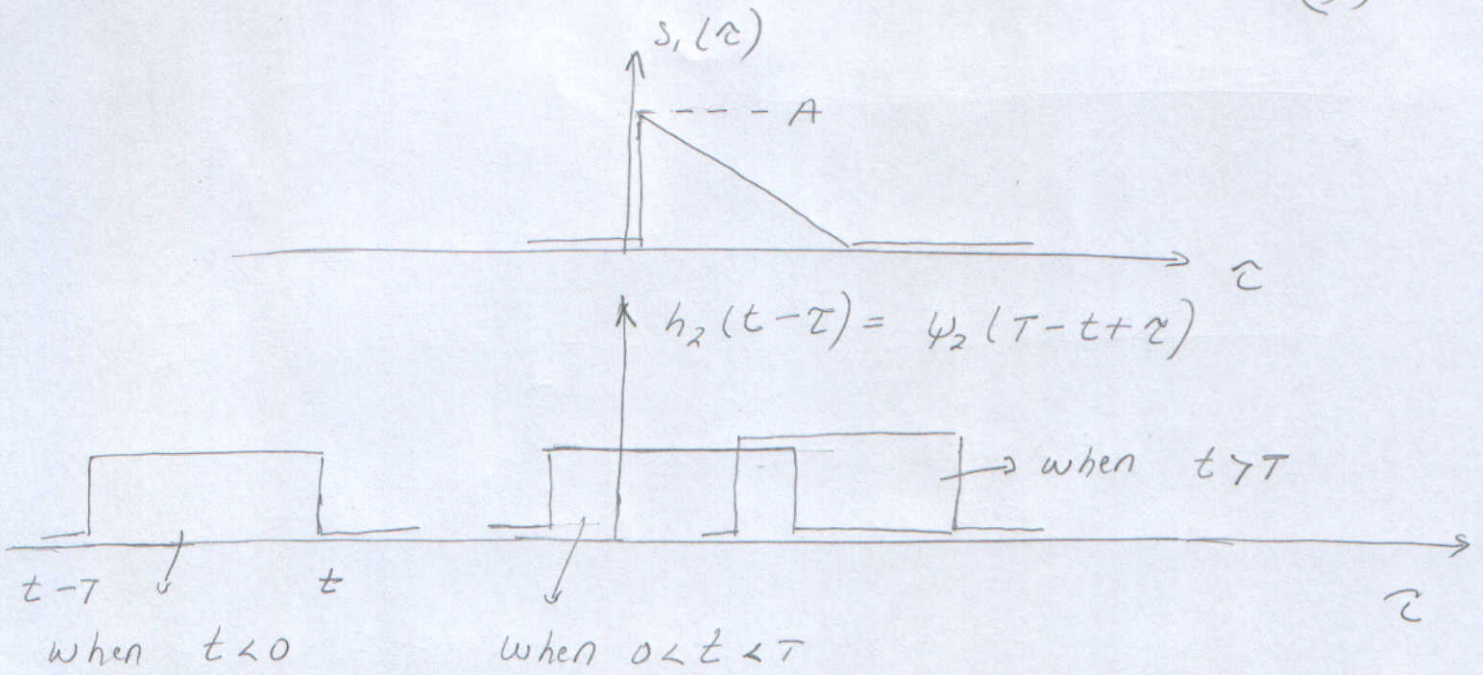
$$= \frac{A}{2} \sqrt{\frac{3}{T}} \left\{ \frac{T^2}{2} - \frac{2(T^3 - T^2t + 0.5T^2t - T^3/3)}{T} \right.$$

$$\left. - T(t-T) + \frac{(t-T)^2}{2} + \frac{2[T^2(t-T) - Tt(t-T) + 0.5t(t-T)^2 - (t-T)^3/3]}{T} \right\}$$

Tests $y_{12s1}(t=T) = \frac{A}{2} \sqrt{\frac{T}{3}}$, $y_{12s1}(t=2T) = 0$: OK

The OPs of MI for the lower branch is shown overleaf

(5)



At $0 < t < T$

$$y_{21s_1}(t) = \int_0^t s_1(\tau) h_2(t-\tau) d\tau$$

$$= \frac{A}{T\sqrt{T}} \left(Tt - \frac{t^2}{2} \right)$$

Tests $y_{21s_1}(t=0) = 0$, $y_{21s_1}(t=T) = \frac{\overbrace{A\sqrt{T}}^{s_{12}}}{2}$

At $T < t < 2T$

$$y_{22s_1}(t) = \int_{t-T}^T s_1(\tau) h_2(t-\tau) d\tau$$

$$= \frac{A}{T\sqrt{T}} \left[\frac{T^2}{2} - T(t-T) + \frac{(t-T)^2}{2} \right]$$

(6)

At $T < t < 2T$

$$y_{2251}(t) = \int_{t-T}^T s_1(\tau) h_2(t-\tau) d\tau$$

$$= \frac{A}{T\sqrt{T}} \left[\frac{T^2}{2} - T(t-T) + \frac{(t-T)^2}{2} \right]$$

Tests $y_{2251}(t=T) = \frac{A\sqrt{T}}{2}$, $y_{2251}(t=2T) = 0$: OK

The correlator will give the OP after integration (not t dependent), it should match the above at $t=T$.

$$y_{151}^c(t) = \int_0^T s_1(t) \psi_1(t) dt$$

$$= \frac{A}{T} \sqrt{\frac{3}{T}} \int_0^T \left(T - 3t + \frac{2t^2}{T} \right) dt$$

$$= \frac{A}{T} \sqrt{\frac{3}{T}} \left(T^2 - \frac{3T^2}{2} + \frac{2T^2}{3} \right) = \frac{A}{2} \sqrt{\frac{T}{3}}$$

Same $\Rightarrow y_{151}(t=T) = y_{R51}^{MF}(t=T)$ of MF

(7)

$$y_{12s1}(t) = \int_0^T s_1(t) \psi_2(t) dt$$

$$= \frac{A}{T} \cdot \frac{1}{\sqrt{T}} \int_0^T (T-t) dt$$

$$= \frac{A}{T^{3/2}} \left[Tt - \frac{t^2}{2} \right]_0^T = \frac{A\sqrt{T}}{2}$$

same as $y_{21s1}(t=T) = y_{22s1}(t=T)$

2. (30 Points) Answer the following questions as **True** or **False**. For the **False** ones give the correct answer or the reason. For the **True** ones justify your answer.

- a) For demodulation of an FM signal, we use PLL : *True ; PLL by extracting the phase term in the modulated signal and differentiating this phase obtains the modulating (message) signal*
- b) Orthogonal signals, when placed on the constellation diagram will have an angle of 180° between them : *False, orthogonal signals have an angle of 90° between them, when shown on the constellation diagram*
- c) FSK and PPM are multidimensional signals, establishing orthogonality by dividing the time domain into non-overlapping slots : *Partially true, this statement applies to PPM, whereas FSK establishes orthogonality by slicing the frequency axis and using the point of orthogonality between Δf and T*
- d) 8 PSK and 8 QAM will need the same channel bandwidth : *True, since Bw is related purely to the dimensionality of signal space (provided both use the same waveform shapes)*
- e) A channel with a rectangular frequency response will have a delta function type time response : *False, an unlimited channel, i.e. $C(f) = 1$, will have $\delta(t)$ type of time response, a channel with rectangular frequency response will have a sinc type time response*